

A Hidden Markov Model Based Recursive Algorithm For Wideband Temporal Spectrum Sensing

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Abstract: Wideband Spectrum Sensing methods can be more efficiently used in order to determine the inactive portions that represent the spectrum holes in the frequency range of a given spectrum. In this paper, we propose, an algorithm that implies Hidden Markov modeling by considering the recursive tree search. The proposed algorithm, considers the dynamic activity of the authorized users within the spectrum. The spectrum holes, smooth in the occurrence of bursting main operator signals can be detected accurately by the planned wideband temporal sensing method. The performance gain, particularly, at the time of small duty-cycle by the licensed operator signals can be demonstrated by the obtained numerical results with the mentioned algorithm.

Keywords: Cognitive radio, spectrum sensing, wideband, hiddenmarkov model.

1 INTRODUCTION

Cognitive Radio assumes a key process, to build the use of the unused Radio Frequency variety organizations without disturbance to the authorized or critical customers [1]. A subjective unapproved patron need to make use of complicated detecting methods, if you want to distinguish the on or off situation of a PU sign inside the variety band. A huge portion of the primary chip away at range detecting focused on a skinny band channel, any region the PU signal is notion to be either dynamic or inert, even a tiny bit instances. The execution of this paper, will be an efficient structure and computationally realistic calculations Well regarded recognition calculations for narrowband detecting carries (1) Energy identification: that calls for no earlier statistics at the channel and it plays inadequately as a result of low sign to-noise ratio (SNR); (2) Matched channel discovery: that calls for priori understanding;(3)Cyclo-stationary detail location: that plays at low SNR and its exhibition is debased at the condition of low PU duty cycle[2]. For band detecting that explicitly reflect on consideration on the changing movement of dynamic PUs. The predominant improvements of our methodology incorporate of (1) utilization of concealed Mark off demonstrating to a divided representation of the variety pass, (2) an algorithmic tree chase for looking and instating range openings to apprehend dynamic aspect flag in the given range band.

The paper can be clarified as: In phase II,we communicate about and check the exhibition of the modern wide band energy identifier range detecting strategy. IN location III, we're building up a framework version for dynamic PU below channel blurring and commotion disabilities. In place IV,we build up our proposed recursive tree search calculation to carry out extensive band fleeting range detecting. In vicinity V, we portray the reproduction that changed into utilized to assessment the proposed calculation with current calculation and gift numerical consequences. Finishing up feedback are given in vicinity VI.

II PERFORMANCE OF WIDEBAND ENERGY DETECTOR

In order to allocate a channel for an SU, when PU is in an idle state, it is necessary for an SU to sense ancomplete band that can vary from the order of 1MHZ-1GHZ along with the determination of the channel boundaries. If the SU cannot leverage someexteriordata, then the wideband spectrum sensing is required. AnSU, can achieve wideband identifying throughinitialization & then return toward multiband otherwise narrowband sensing through usual procedure. For this purpose, we are considering the GMSK signals, as they exhibit gradual sloping band edges with limits in the borders amongst signals. Not only GMSK consume radicallynot the same band boundaries, then GMSK is prevalent in modern wireless morals such such asGSM(GMSK). We assume that a channel can take on one of two states: (1) an work-shy state, in which the PU do not convey, that is denoted by 0. (2)an energetic state, in which the PU communicate, that is denoted by 1.Aimed by a agreed network, testable state possibilities are represented as π_0 for an idle state of PU and π_1 for an active state of the PU, that corresponds to the responsibility rotation of the network which is detailed by way of a percentage rate.

2.1 Wideband Energy Detectors

The identifying marks of a frequency-domain vitality finder aimed at GMSK signals can be shown in the figure1, where10dB is considered as the SNR and the greatness of the PSD price losses by the duty cycle for the bursting

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signals. In the figure1,the spectrum holes can be viewed as the shaded areas. Extreme grip task is used than the be an average of PSD settings aimed at detecting dynamic PU's .As this extreme grip vitality detector be able to actually product in upper probability of wrong alarm by way of opinion distances are enlarged outstanding towards increase probability of an unusually great noise power throughout the sensing interval. Because of these drawbacks, we are using the wideband sensing algorithm.

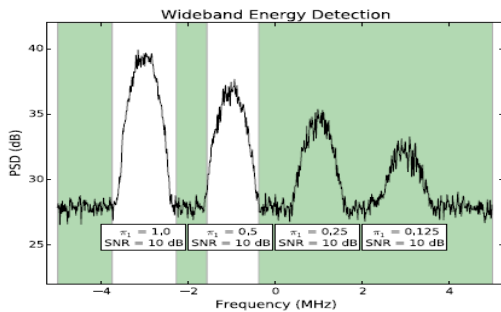


Fig. 1. Sensing results of wideband energy detector with duty cycles of 100%,50%,25%,and 12.5% along with 10dB SNR.

III System Model

In this section, we are developing a system ideal on behalf of the dynamic PU beneath network fading in addition to noise losses. Here, we are considering a spectrum band of total bandwidth of BHz with unknown number of PU signals of 0 to 100 percent temporal duty cycles. The middle frequency of each PU signal has f_c Hz and the bandwidth of WHz. Let us consider i th PU on the channel is level Rayleigh fading through parameters $\sigma_{f,i}$ that mutual through the Additive White Gaussian Noise(AWGN) definite through circularly symmetric complex regular circulation $C(0, \sigma_{n,i}^2)$. The received signal's mean SNR on channel i , when the PU is conducting is $SNR_i = \sigma_{f,i}^2 / \sigma_{n,i}^2 (1)$

3.1 PUTraffic Model

The two states of the PU , be either wavelength otherwise idle at several assumed time, where the idle state of the i th PU is signified through $X_i=0$, where the PU is not conveying . The k th PU state is denoted by $X_{i,k}$. Every PU is modeled through a separate-time Markov chain by way of transition matrix G_i as primary distribution v_i , definite as

$$G_i = [g_{i,ab}: a, b \in \{0,1\}] \tag{2}$$

$$g_{i,ab} = P(X_{i,k} = a, X_{i,k+1} = b) \tag{3}$$

$$v_i,0 = P(X_{i,1} = 0), v_i,1 = P(X_{i,1} = 1) \tag{4}$$

3.2 Cognitive Receiver Model

3.2.1 Received Wideband Signal

A spreading PU resolve produce a band pass signal t_i,k . The transmitting indication aimed at PU i by every period k remains

$$t_i,k = t_i,k.1 \{X_{i,k}\} \tag{5}$$

Where 1A remains the pointer job happening the incident or situation A. The i th PU signal remains grown at period K through a disappearing indicator $f_i,k \sim C(0, \sigma_{f,i}^2)$. Everything M PU signals remain conventional at the same

time & extra towards sound signal $n_k \sim C(0, \sigma_{n,i}^2)$. The received wideband signal is signified through a arrangement of examples $Z_{wbn} = \{Z_{wbn,1}, \dots, Z_{wbn,n}\}$, where $Z_{wbn,k}$, the k th I-Q sample since the wideband network, remains well-defined by way of $Z_{wbn,k} = \sum_{i=1}^M t_i,k f_i,k + n_k \tag{6)$

3.2.2 Channelized Received Signal

In this, the conventional wideband signal is divided into J narrowband sub band channels by an SU. Primarily, this separation obligation remain completed randomly but then again later wideband identifying, the usual of sub channels must define completely the PU data, as fine as range holes among PU signals. The j th sub band is signified by Z_j .

3.2.3 Energy Detected Signal

The channelized narrowband indication remain managed by an be around power indicator, which estimations the authority of both exemplar so medians N_{avg} sample composed. The conventional power in sub channel j is represented through Y_j where its energy estimates are, $Y_j = \{y_{j,1}, \dots, y_{j,n}\}$ and the k th instance happening the power finding arrangement $y_{j,k}$ can be then defined.

IV Recursive Algorithm for Wideband Temporal Sensing

In this segment, we suggest a method that spreads narrowband temporal identifying methods towards the wideband scenario. The below algorithms 1 and 2 represents the suggested recursive wideband time-based identifying structure.

4.1 Algorithm -1

This algorithm is used on behalf of Wideband historical sensing and remains explained in the below manner.

Step1: First, we sense the center frequency (f_c), bandwidth (W), desired resolution (W_r) and the received wideband signal (Z_{wbn}), by using the function R_{sense} , $R_{sense}(f_c, W, W_r, Z_{wbn})$; Step2: Next, it checks if $W > W_r$, then it calculates the

$$L_{lo} = R_{sense}(f_c - W/2, W/2, W_r, Z_{wbn});$$

$$L_{hi} = R_{sense}(f_c + W/2, W/2, W_r, Z_{wbn});$$

Step3: After calculating the L_{lo} and L_{hi} , then it checks whether the L_{lo} and L_{hi} are empty or not.

Step4: In case, if they are not empty, then it performs

$$L = \text{AggregateCh}(L_{hi}, L_{lo}, Z_{wbn});$$

Step5: If not, they goes to L , which is the empty list.

Step6: Then if $W > W_r$ is false, it first calculates the discrete taps $h(n)$ by

$$h(n) = \text{LPF}(W, N_t);$$

where N_t is the number of filter taps for the channel selecting LPF.

Step7: In this step, the $h(n)$ is decimated as $Dec = \text{Floor}(W_0/W)$;

Step8: In order to perform channelization,

$$\text{we use } z_n = \text{DDC}(Z_{wbn}, f_c, h(n), dec);$$

Step9: After performing channelization, we are performing energy detection based on the processed received power samples.

$$y_n = \text{EnergyDet}(Z_n);$$

Step10: In the narrowband channel with processed received power samples, y_n , the parameter of the PU is obtained by the $BaumEst(y_n) = (v, G, \mu, R, x^n)$;

Step11: The stationary state distribution corresponding to the transition matrix G , can be computed by, $\pi = \text{StatDistr}(G)$;
 Step12: After computing, then the channel usability is done by considering, if $\pi_1 > \pi_{\min,1}$
 Then, $L = \text{list}$ with single entry (f_c, W, z_n, x^n) ;
 Step13: If $\pi_1 > \pi_{\min,1}$ is false, it return to $L = \text{empty list}$;
 Step14: Then return to the estimated channel parameter L .

4.2 Algorithm-2

The below algorithm describes, whether the two adjacent holes should be combined and combining of the two channels and so on.

Step1: Initially, it is checked whether two adjacent holes are combined or not by using the equation,

$\text{AggregateCh}(L_{hi}, L_{lo}, z_{wbn})$.

Step2: Then the lowest frequency narrowband channel is selected by $(f_{lo}, c, W_{lo}, z_{lon}, x_{lo}^n)$ and the highest frequency narrowband channel is selected by $(f_{hi}, c, W_{hi}, z_{hin}, x_{hi}^n)$.

Step3: After the selection of the channels, the function $\text{CombineLists}(L_{lo}, L_{hi})$ is used to merge two lists of estimated channel parameters into a single list and sorts the list in decreasing order of center frequency.

Step4: Next, if $f_{hi}, c - W_{hi}/2 = f_{lo}, c + W_{lo}/2$ then the modified correlation metric can be computed by, $\rho = \text{correlate}(n, z_{lon}, x_{lo}^n, z_{hin}, x_{hi}^n)$;

Step5: In case, if $\rho > \rho_{\min}$, then remove $(f_{lo}, c, W_{lo}, z_{lon}, x_{lo}^n)$ and $(f_{hi}, c, W_{hi}, z_{hin}, x_{hi}^n)$ from L .

Step6: Then these are low pass filtered by the discrete filter taps $h(n)$ and decimated by dec .

Step7: After decimation, the center frequency (f_c) and the received power samples (z_n) are derived.

Step8: The MAP decisions from the two channels are combined by using $\text{merge}(x_{lo}^n, x_{hi}^n)$;

Step9: By adding them, as shown in the below expression, to L .

Add $(f_c, W_{lo} + W_{hi}, z_n, x^n)$;

Step10: Finally, return to L .

In the section V, we can see the final simulation results after using the above two algorithms.

V Simulation and Results

In this segment, we explain the consequences that are attained after simulation.

5.1 Comparison of Techniques By Quantitative Manner:

A ROC curve produced through the replication represents the quantitative results. The possibility of true notice is plotted against the probability of false detect. These curves signify the normal detector presentation over several random wideband imprisonments consuming the same variation, duty cycle also SNR. In the wideband capture result, it can be clearly observed that the detector enactment degrades as PU duty cycle reductions. The presentation of the GMSK energy detector performance is shown in fig 2, where the detector performance degrades as PU duty cycle is decreased. From the figure 3, we can state that the presentation of the wideband temporal energy indicator does not show any performance degradation by reducing the duty cycles.

5.2 Simulation Results: cfar Curves for Wideband Temporal Detector For GMSK Signals

In this simulation, at all the tested SNR & power detected windows, the false alarm and true positive rates are taken. At a constant false alarm rate (CFAR) of 0.01, many ROC bends are imposed on a single plot. The increase in the sensitivity is due to the increase in the energy detection window for GMSK signals in figure 4. The detector performance is degraded, as the vitality detector length grows along with the probability that samples the idle also busy cycles that are averaged together.

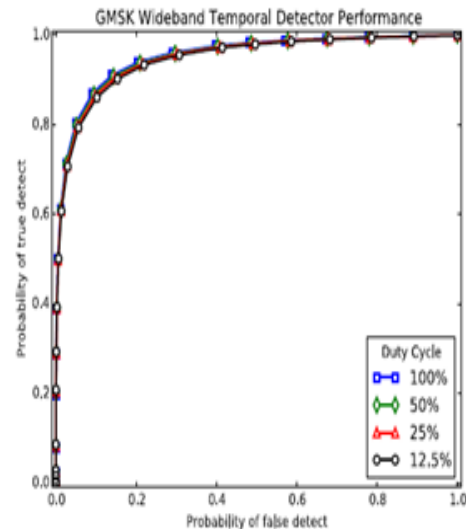


Fig.2. ROC curve on behalf of wideband power detector for GMSK signals through 10dB SNR & 100%, 50%, 25% & 12.5% duty cycles.

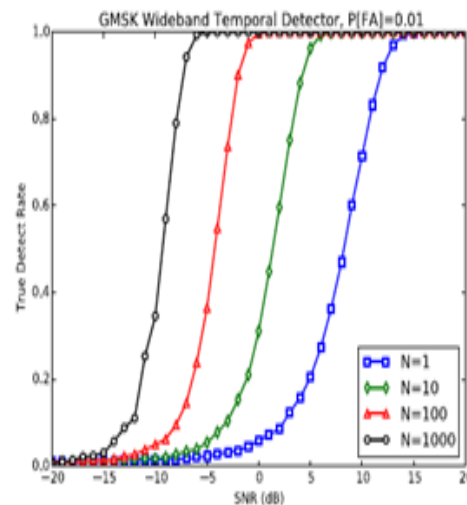


Fig3:- ROC curve on behalf of wideband temporal range indicator for GMSK signals through 10dB SNR & 100%, 50%, 25% and 12.5% duty cycles

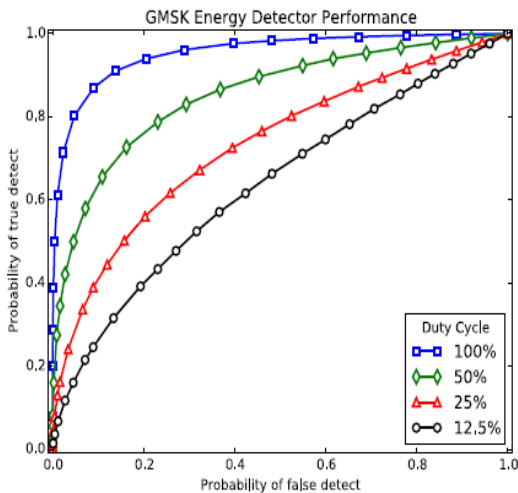


Fig.4. CFAR curves for wideband temporal spectrum detector on behalf of GMSK signals through duty cycle of 12.5%, SNR reaching from -20 to 20dB also energy detection window of 1, 10, 100 or 1000 samples.

VI Conclusion

In this paper, we consume shown that the existing wideband energy detector performance gain can be increased even in the situation of bursting indications through stumpy duty cycle, by considering the wideband temporal spectrum sensing for GMSK signals. By using the proposed method, the changing activity of the primary users is specified correctly. The method is simple in direction to perceive the spectrum holes most efficiently. The recursive algorithm is proved to be much robust in the case of the bursting signals for GMSK signals. The entire proposed method is simulated by using the MATLAB software and achieved the required numerical results more accurately.

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